

Channel Models for Medical Implant Communication

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Received: 31 October 2010/Accepted: 2 November 2010/Published online: 9 December 2010
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Abstract Information regarding the propagation media is typically gathered by conducting physical experiments, measuring and processing the corresponding data to obtain channel characteristics. When this propagation media is human body, for example in case of medical implants, then this approach might not be practical. In this paper, an immersive visualization environment is presented, which is used as a scientific instrument that gives us the ability to observe RF propagation from medical implants inside a human body. This virtual environment allows for more natural interaction between experts with different backgrounds, such as engineering and medical sciences. Here, we show how this platform has been used to determine channel models for medical implant communication systems.

Keywords Channel model · Implant communication system · Immersive visualization system · Body area networks

1 Introduction

Body Area Networks (BAN) consist of RF-enabled wearable and implantable sensory nodes for short range reliable communication in or around a human body. This technology is poised to be a promising interdisciplinary technology with novel uses in pervasive information technology. Pacemakers are the oldest and most successful example of implantable medical devices. Among other applications for such devices, we can point to smart pills for precision drug delivery, glucose monitors and eye pressure sensing systems [1–4]. Similarly, wearable sensor nodes offer an attractive set of applications such as medical/physiological monitoring (e.g. electrocardiogram, temperature, respiration, heart rate, blood pressure), disability assistance and human performance management. Integration of BAN with the existing information infrastructure will create a truly pervasive environment for many of these critical applications with great impact on improving the quality of life. Some recent advances in microelectronics indicate that the technology to achieve ultra-small and ultra-low power devices for these applications is within reach [5, 6]. However, numerous regulatory and technical challenges including efficient transceiver design, reliability, biocompatibility, size, cost, energy source, sensing/actuator technology, privacy and security issues still need to be resolved [7, 8].

In 1999, the Federal Communications Commission (FCC) allocated the frequencies in the 402–405 MHz range to be used for Medical Implant Communication Services

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(MICS). This is an ultra-low power, unlicensed mobile radio service for transmitting data in support of diagnostic or therapeutic functions associated with implanted medical devices. The MICS permits individuals and medical practitioners to utilize ultra-low power medical implant devices, such as cardiac pacemakers and defibrillators without causing interference to other users of the electromagnetic radio spectrum. Among primary reasons for selecting these frequencies, one can point to better propagation characteristics for medical implants, reasonable sized antennas and worldwide availability. An international radio standard for MICS will allow patients with implantable devices to obtain necessary care worldwide.

Knowledge of RF propagation for implantable devices will assist RF engineers to optimize physical layer design and therefore, achieve better performance. Such information is typically gathered by conducting physical experiments, measuring and processing the corresponding data to obtain channel characteristics. In case of medical implants, this could be extremely difficult if at all possible. In this paper, we present a sophisticated and innovative 3D virtual reality simulation platform in order to study electromagnetic propagation from medical implants. Communication with an implanted device can be done from any direction inside or outside the body due to various body postures and human motion. Consequently, a true 3D environment is needed to better capture, visualize and understand RF propagation from/to implants. In the following sections, we describe such a platform and show how it was used to extract a simple statistical path loss model for MICS channels.

This paper is organized as follows. Section 2 will describe the immersive 3D platform that we have constructed to study RF propagation from radio implants. The simulated scenarios and description of our approach are provided in Sect. 3. Then, the statistical path loss model is described in Sect. 4, and finally concluding remarks and future plans are expressed in Sect. 5.

2 A 3D Immersive Platform for Medical Implants

Block diagram of our simulation system is shown in Fig. 1. The main components of this system include: a three-dimensional human body model, the propagation engine which is a three-dimensional full-wave electromagnetic field simulator (i.e. HFSS¹), the 3D immersive and visualization platform, and finally an implantable (or body surface antenna).

¹ HFSS is registered trademark of ANSYS Inc. The HFSS has been used in this research to foster understanding. Such identification does not imply recommendation or endorsement by the National Institute of Standards and Technology, nor does it imply that this product is necessarily the best available for the purpose.

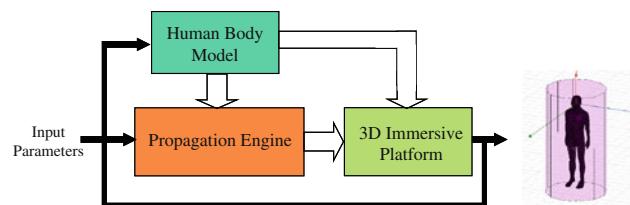


Fig. 1 System block diagram

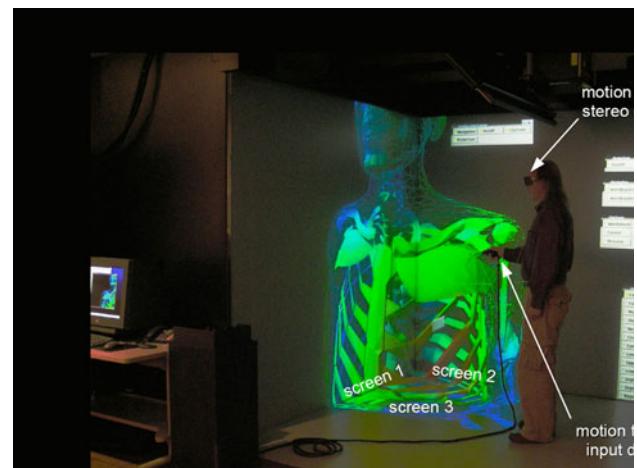


Fig. 2 A user in the NIST immersive visualization environment

The 3D human body model includes frequency dependent dielectric properties of 300+ parts in a male human body. These properties are also user-definable if custom changes or modifications are desired. The human body model has a resolution of 2 mm. The HFSS propagation engine enables us to compute a variety of different electromagnetic quantities such as the magnitude of electric and magnetic fields and Specific Absorption Rate (SAR).

The 3D immersive platform as shown in Fig. 2 includes several important components: three orthogonal screens that provide the visual display, the motion tracked stereoscope glasses, and the hand-held motion tracked input device. The screens are large projection video displays that are placed edge-to-edge in a corner configuration. These three screens are used to display a single three-dimensional (3D) stereo scene. The scene is updated based on the position of the user as determined by the motion tracker. This allows the system to present to the user a 3D virtual world within which the user can move and interact with the virtual objects. The main interaction device is a hand-held three button motion-tracked wand with a joystick.

This virtual environment allows for more natural interaction between experts with different backgrounds such as engineering and medical sciences. The user can look at data representations at any scale and position, move through data, change orientation, and control the elements

of the virtual world using a variety of interactive measurement and analysis techniques [9]. All of these capabilities are extremely useful when studying RF propagation to/from medical implants.

The final component of our system is the implant antenna. The operating environment for an implant antenna is quite different from the traditional free space communication. Designing an efficient antenna for implantable devices is an essential requirement for reliable MICS operation [10]. The dimension of the antenna must be very small and it should be long term biocompatible [11]. The antenna must also be electrically insulated from the body to avoid possible short circuits [12].

Figures 3 and 4 display the implant antenna used in our simulations. The antenna is composed of a single metallic layer and is printed on a side of a D51 (NTK) substrate with dielectric constant of $\epsilon_r = 30$, loss tangent of $\tan \theta = 0.000038$, and thickness of 1 mm. The metallic layer is copper with 0.036 mm thickness. The dimension of the antenna is $8.2 \times 8.1 \times 1$ mm which is quite appropriate for some medical applications. The metallic layer is covered by RH-5 substrate with dielectric constant of $\epsilon_r = 1.0006$, loss tangent of $\tan \theta = 0$, and thickness of 1 mm. The simulated return loss of this antenna is shown in Fig. 5. Good impedance matching in the MICS frequency band (i.e. 402–405 MHz) is observed. Further details on the design and performance of this antenna can be found in [13, 14].

Input parameters to our system include: antenna position, antenna orientation, operating frequency, transmit power, resolution, range and the choice of the desired

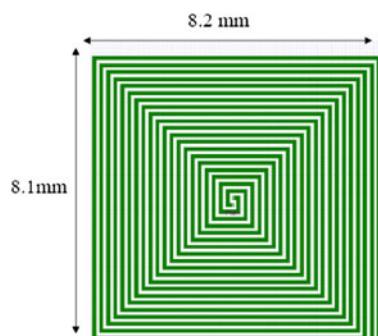


Fig. 3 Front view of the implant antenna

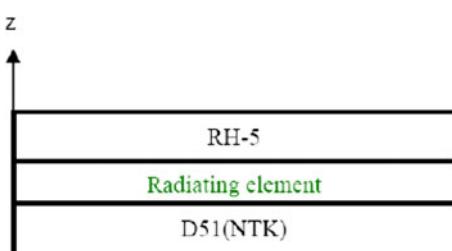


Fig. 4 Side view of the implant antenna

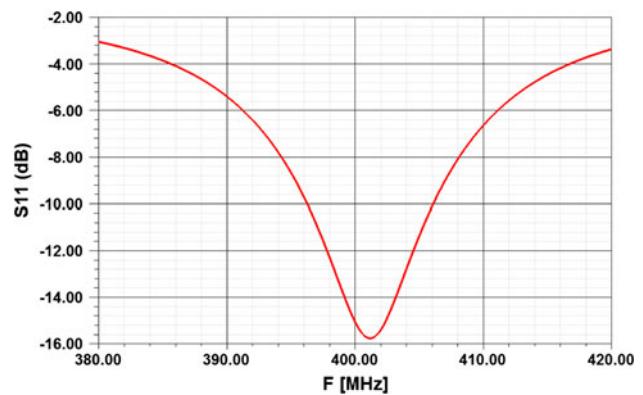


Fig. 5 Return loss of the implant antenna

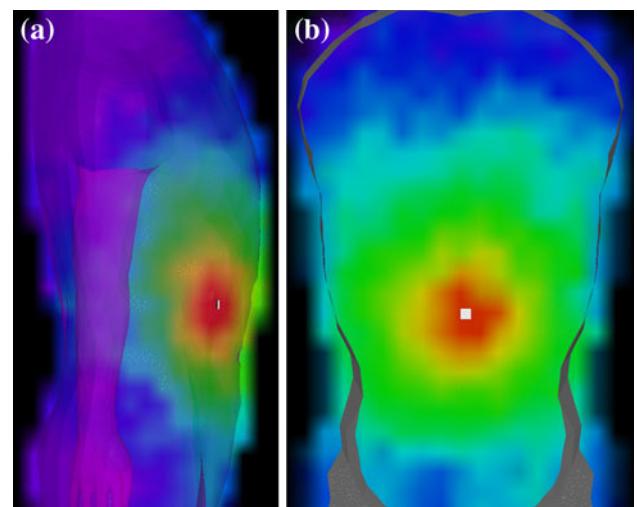


Fig. 6 Sample output image of the immersive system. **a** Side view. **b** Front view

output parameters. The operating frequency in this study has been chosen to be 403.5 MHz which is the mid-point of the MICS frequency band. Resolution of 4 mm has been selected to run the simulation and a range (i.e. distance) of 50 cm from the transmitting antenna has been considered.

Figure 6 displays sample output images of our immersive system when an antenna is located in the stomach (e.g. side and front views). The color map shows the signal strength on a 2D plane which is perpendicular (Fig. 6a) or parallel (Fig. 6b) to the antenna plane.

3 Simulation Scenarios

Simulations have been performed for four near-surface implants and two deep-tissue implants applications in a typical male body. The near-surface scenarios include applications such as Implantable Cardioverter-Defibrillator (ICD) and Pacemaker (located below the left pectoral

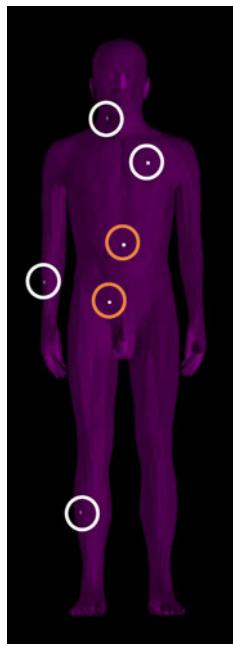


Fig. 7 Various transmitter locations for obtaining signal strength

muscle), Vagus Nerve Stimulation (Right Neck and Shoulder) and two Motion Sensor applications located in right hand and right leg. These locations are highlighted by the white circles in Fig. 7.

The deep tissue implant scenarios consider endoscopy capsule applications for upper stomach (95 mm below body surface) and lower stomach (118 mm below body surface) as highlighted by the red circles in Fig. 7.

For each scenario (i.e. TX location), the received power was calculated for a grid of points within a cylinder area around the body. Then, the resulting data was partitioned into three sets: in-body to in-body, in-body to body surface, and in-body to out-body propagation sets. The in-body to in-body set includes all of the sample points that completely reside inside the body. Likewise, the in-body to body surface set includes all points that reside within a definable distance (i.e. 2 mm, 10 mm, and 20 mm) from the body surface; and finally the in-body to out-body propagation set distinguishes all of the points that reside further away from the body surface.

4 Path Loss Model

Path loss at a distance d from the transmitting antenna, in our calculation, is defined as:

$$PL(d) = \frac{G_R P_T}{P_R(d)}$$

where P_T is the transmit power, P_R denotes the received power and G_R is the receiver antenna gain. Therefore, as

defined in the above equation, the path loss would include the transmitter antenna gain. This is usually not the case for channel models corresponding to most wireless systems, but for MICS, the transmitting antenna is considered to be part of the channel [15]. The path loss in dB at distance d can be statistically modeled by the following equation:

$$PL(d) = PL(d_0) + 10n \log_{10}(d/d_0) + S \quad d \geq d_0$$

where d_0 is the reference distance (i.e. 50 mm), and n is the path loss exponent which heavily depends on the environment where RF signal is propagating through. For, example, it is well known that for free space $n = 2$. Human body is an extremely lossy environment; therefore, much higher value for the path loss exponent is expected. S is the random scatter around the mean and represents deviation in dB caused by different body materials (e.g. bone, muscle, fat, etc.) and the antenna gain in different directions.

We first model the path loss by finding the values for $PL(d_0)$, n and S for deep-tissue versus near-surface implants separately. This will lead to four propagation scenarios i.e. deep-tissue implant-to-implant, near-surface implant-to-implant, deep-tissue implant to body surface and finally near-surface implant to body surface. Next, we combine the deep-tissue and near-surface implant scenarios into one set of data and only consider the propagation channel for the following two cases: implant-to-implant and implant to body surface. This should give us more insight into possible differences between communication channels related to the deep tissue and near surface implants.

Figure 8 shows the scatter plot for the path loss as a function of TX-RX separation for deep tissue implant-to-implant scenarios. The mean value of the random path loss has been displayed by a solid line. This is obtained by

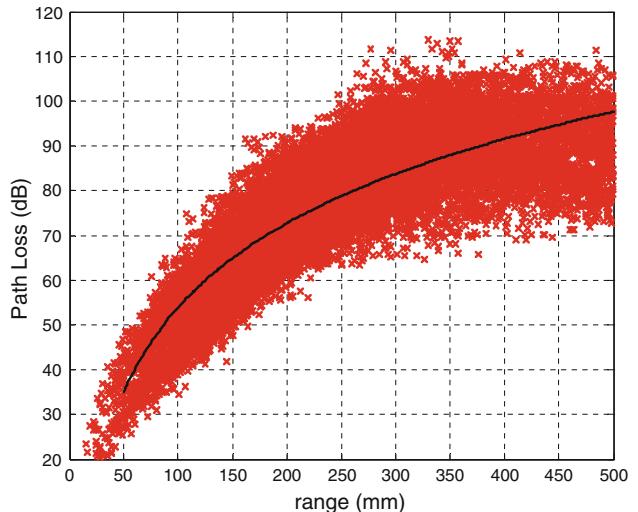


Fig. 8 Scatter plot of the path loss versus distance for deep tissue implant to another implant

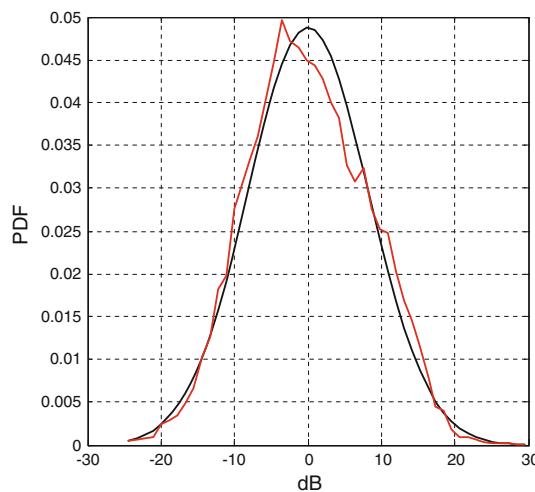


Fig. 9 Distribution of the shadow fading for deep tissue implant to another implant

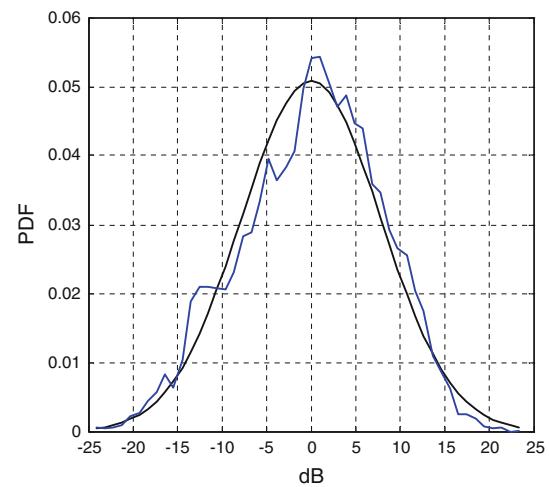


Fig. 11 Distribution of the shadow fading for deep tissue implant to body surface (within 20 mm of the skin)

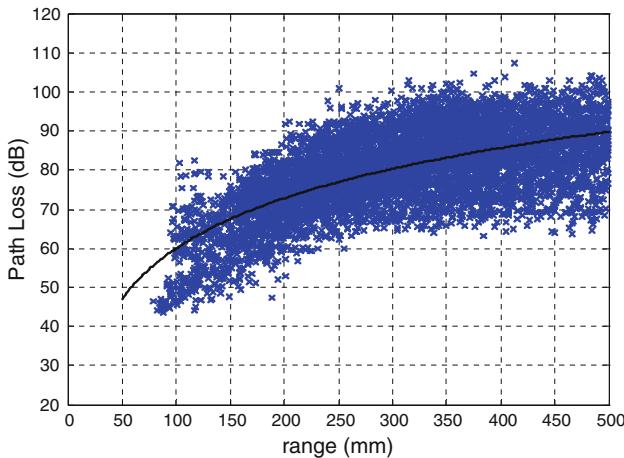


Fig. 10 Scatter plot of the path loss versus distance for deep tissue implant to body surface (within 20 mm of the skin)

fitting a least squares linear regression line through the scatter of measured path loss sample points in dB such that the root mean square deviation of sample points about the regression line is minimized. Random shadowing effects of the channel occur where the TX–RX separation is the same, but have different directions or positions with respect to each other. As shown in Fig. 9, this random variable has a normal distribution with zero mean and standard deviation σ_s i.e. $S \sim N(0, \sigma_s^2)$.

Similarly, Figs. 10 and 11 represent the path loss scatter plot and probability density function of the shadow fading random variable for the deep tissue implant to body surface scenarios. A distance of up to 20 mm directly from the body surface has been considered in the definition of the body surface sample points. We also tried a separation distance of 2 and 10 mm and observed that the derived path

Table 1 Parameters for the statistical path loss model for the case of deep tissue implant

	PL(d_0) (dB)	n	σ_s (dB)
Deep tissue implant to another implant	35.04	6.26	8.18
Deep tissue implant to body surface	47.14	4.26	7.85

loss parameters did not vary much. One should keep in mind that layers of clothing could cause additional loss to the signal.

Table 1 summarizes the extracted parameters of the statistical path loss models for the above scenarios.

Scatter plots and shadow fading distributions corresponding to the near-surface implant scenarios have been provided in Figs. 12, 13, 14, and 15.

Table 2 summarizes the extracted parameters of the statistical path loss models for the near surface implant scenario.

Deriving a channel model for communication between an implant and another node that resides further away from the body is more challenging. In the simple case of an environment where there are no objects or obstacles, a free space path loss can be added to the above models to account for the additional loss that the implant signal will go through once it leaves the body. On the other hand, if there are objects adjacent to the body in the surrounding environment, further loss could occur and the model needs to be adjusted accordingly. With appropriate assumptions this channel could be considered as an aggregate of two concatenated channels: implant to body surface and body surface to the external node channels. Further analysis and measurement data is needed to investigate this channel and obtain the appropriate statistical models.

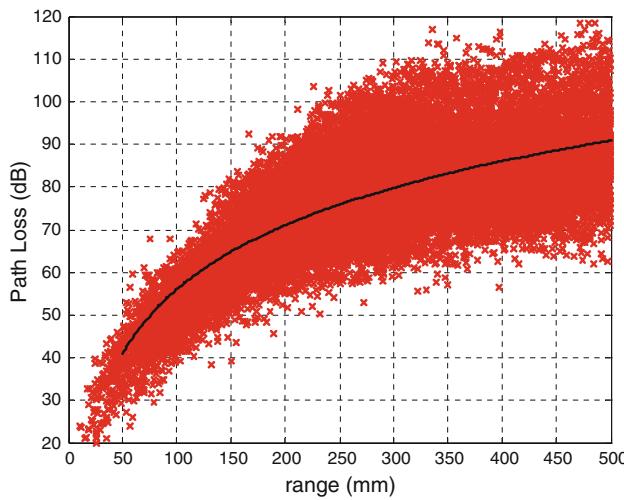


Fig. 12 Scatter plot of the path loss versus distance for near surface implant to another implant

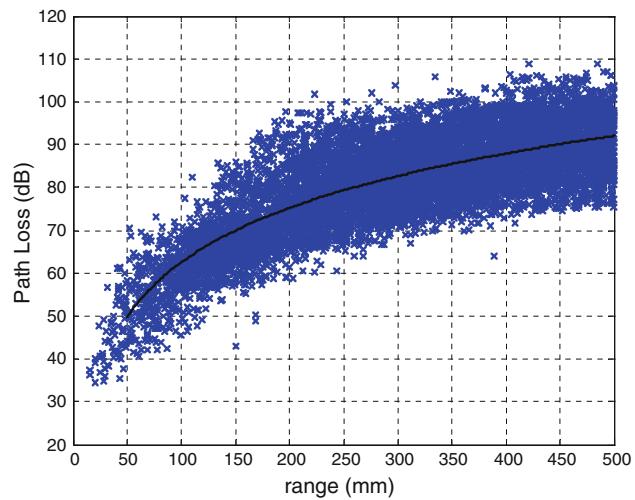


Fig. 14 Scatter plot of the path loss versus distance for near surface implant to body surface (within 20 mm of the skin)

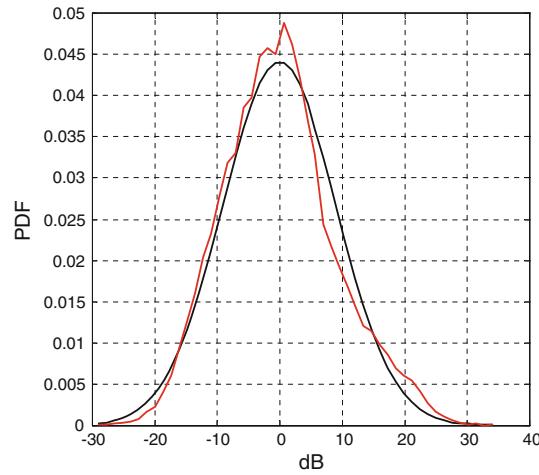


Fig. 13 Distribution of the shadow fading for near surface implant to another implant

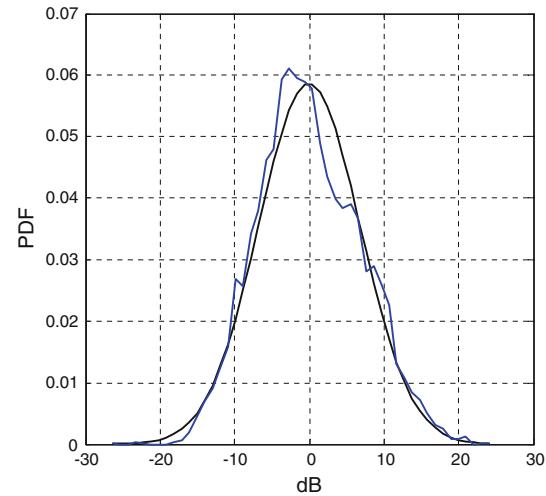


Fig. 15 Distribution of the shadow fading for near surface implant to body surface (within 20 mm of the skin)

5 Conclusion

We have presented an immersive visualization environment to conduct research in order to characterize RF propagation from medical implants. Extensive simulations have been performed to obtain a statistical path loss model for MICS channels. The model is based on four near surface and two deep tissue implant applications in a typical male human body. The authors recognize the fact that the extracted parameters for the statistical model are based on the simulation data; therefore, upon availability of measurement data from physical experiments, the results outlined in this paper should be further validated. However, no such reference data set is currently available. Along this line, efforts are underway to conduct and obtain body surface measurements which can be emulated in the

Table 2 Parameters for the statistical path loss model for the case of near surface implant

	$PL(d_0)$ (dB)	n	σ_s (dB)
Near surface implant to another implant	40.94	4.99	9.05
Near surface implant to body surface	49.81	4.22	6.81

immersive system for cross-verification purposes. The path loss models obtained in this study have been adopted by the IEEE802.15 task group TG6 on body area networks. The accuracy and applicability of such models requires further validation and investigation by scientists and engineers.

In general, studying a specific medical implant application with custom made antennas is also possible with this platform. More in-depth research on this subject is

undoubtedly required to further understand the characteristics of radio frequency propagation from medical implants. The authors hope that the virtual reality environment introduced here would create a flexible platform where more efficient collaboration between engineers and medical experts would become possible.

Acknowledgment The authors would like to thank Mr. Steven Satterfield from the high performance computing and visualization group of the Applied and Computational Mathematics Division, ITL, for his assistance in the system setup and his ongoing contribution in the development of tools for the NIST immersive visualization system.

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Kiyoshi Hamaguchi received the B.S. and M.S. degrees in electrical engineering from Science University of Tokyo in 1989 and 1991, respectively. He also received the Ph.D. degree in electrical engineering from Osaka University in 2000. Since 1993 he has been with the National Institute of Information and Communications Technology (NICT), Japan, where he has been engaged in research and development on wireless telecommunication systems. From 2002 to 2003, he was a visiting researcher at University of Southampton, U.K. He received the Young Engineer Award from IEICE in 1997, the Young Scientist Award from Ministry of Education, Culture, Sports, Science and Technology of Japan in 2006 and the Radio Achievement Award from the Association of Radio Industries and Businesses (ARIB) in 2010. He is currently a group leader of Medical ICT group in NICT.